

STUDY OF AN ADD-DROP FILTER USING A SINGLE BRAGG COUPLER

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ABSTRACT

An add-drop filter that consists of a planar directional coupler with a single Bragg grating in one of its arms was studied. The calculation was based on coupled-mode equations in the form of a matrix equation by considering the studied structure. From the matrix equation, the propagation factors as eigenvalues of the eigenmodes was obtained, while the spatial structures of the eigenmodes are determined by the eigenvectors of the coupling matrix. By use of boundary conditions in individual waveguides, the drop output, return loss and total transmission, respectively, was calculated at the Bragg wavelength of 1550 nm. The results satisfy the requirements for Wavelength Division Multiplexing (WDM) applications.

Keywords : Bragg grating, add-drop filter, coupled mode equation, matrix equation

STUDI SUATU FILTER ADD-DROP DENGAN KOPLER BRAGG TUNGGAL

ABSTRAK

Telah dikaji suatu filter add-drop yang terdiri dari suatu papah directional coupler, dimana salah satu lengannya dibuat suatu Bragg tunggal. Perhitungan didasarkan pada persamaan moda terganggu yang dibuat dalam bentuk persamaan matriks, dengan mempertimbangkan struktur yang dikaji. Dari persamaan matriks, diperoleh faktor-faktor propagasi sebagai nilai eigen dari modulus-eigen, sedangkan struktur ruang dari modulus-eigen ditentukan dengan vektor-vektor eigen dari matriks terganggu. Output dari filter add-drop, kerugian balik dan transmisi total dihitung dengan menggunakan syarat-syarat batas dalam masing-masing pandu gelombang pada panjang gelombang Bragg 1550 nm. Hasil yang diperoleh memenuhi persyaratan untuk aplikasi Wavelength Division Multiplexing (WDM) dalam sistem telekomunikasi.

Kata kunci : grating Bragg, filter add-drop, persamaan moda terganggu, persamaan matriks

INTRODUCTION

A fiber-optic system employing wavelength-division multiplexing needs add-drop filters. The simplest implementation of such filters is a four-port device designed to add and to drop a single channel. It generally consists of multi-channel input and output ports, which together form a main signal bus, and single-channel drop and add ports. Numerous all-waveguide add-drop filter schemes have been demonstrated. An add-drop using a directional coupler with a single Bragg grating in one of its arms has

been proposed and experimentally realized by Dong et al. (1996). Motivated by this filter, Erdogan (1998) has studied theoretically an asymmetric planar Bragg coupler. Instead of analyzing the coupled-mode equations, his calculations were based on the role of the waveguides' thickness, with difference V -numbers of the waveguides and their separation on the filters' performances. On the other hand, by employing coupled-mode equations, Orlov et al. (1997) analyzed theoretically the performance of an add-drop filter using a directional coupler with the same waveguides that consist of the same Bragg gratings. However, such filter is not easy to realize experimentally. Therefore, following the way of Orlov et al.(1997), we study the add-drop filter using a single Bragg coupler, where the two planar waveguides have the same V -numbers and lengths.

In this study, we use the same coupled-mode equations with those of Orlov et al. (1997) but with a single uniform Bragg grating without apodization. The coupled-mode equations are presented in the form of a matrix equation, and from its' coupling matrix we solve the propagation factors as eigenvalues of the eigenmodes, while the spatial structures of the eigenmodes are determined by the eigenvectors of the coupling matrix. By using the propagation factors, eigenvectors, and boundary conditions in individual fibers, we obtain the eigenmodes. Finally, we calculate and analyze device parameters such as drop output, return loss and total return loss, respectively.

METHOD

An add-drop filter using a directional coupler with a single Bragg grating in one of its arms was studied theoretically by use of coupled mode equations in form of

matrix equation. The outputs of the structure were obtained by solving those equations numerically.

We consider the total TE electric field $E_y(x,z,t)$ in the structure shown by Figure 1 as follows:

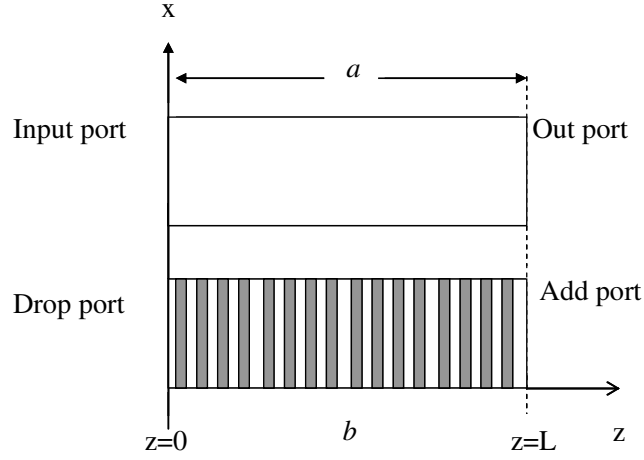


Figure 1. Structure of the studied add–drop filter.

$$E_y(x, z, t) = \left\{ \left[A_1(z)e^{-i\beta_a z} + A_2(z)e^{i\beta_a z} \right] e_{y_a}(x) + \left[B_1(z)e^{-i\beta_b z} + B_2(z)e^{i\beta_b z} \right] e_{y_b}(x) e^{i\omega t} \right\} \quad (1)$$

where $A_1(z)$ and $A_2(z)$ describe the forward and backward propagating modes in waveguide a, while $B_1(z)$ and $B_2(z)$ describe forward and backward propagating modes in waveguide b. This field satisfies the wave equation

$$\nabla^2 E_y + (n^2 \omega^2 / c^2) E_y = 0 \quad (2)$$

where the refractive index n^2 is denoted as

$$n^2(x, z) = n_s^2 + \Delta n_a^2(x) + \Delta n_b^2(x, z) \quad (3a)$$

and

$$\Delta n_a^2 = 2n_s n_{1a}(x); \Delta n_b^2 = 2n_s n_{1b}(x) \cos Gz \quad (3b)$$

Here, n_s is the substrate refractive index, $G = 2\pi/\Lambda$, Λ is the period of the Bragg grating and the Bragg wavelength $\lambda_B = 2n_s \Lambda$. By substituting Equation (1) into Equation (2), and noting that each waveguide satisfies the following wave equation:

$$\frac{\partial^2 e_{y\alpha}}{\partial x^2} + \left[(n_s^2 + \Delta n_\alpha^2) \frac{\omega^2}{c^2} - \beta_\alpha^2 \right] e_{y\alpha} = 0; \alpha = a, b \quad (4)$$

and using slow varying amplitudes and spatial averaging, one will find the following coupled-mode equations (Yariv and Yeh, 1984) :

$$\begin{aligned} \frac{\partial A_1}{\partial z} &= -i\kappa_{aa} A_2 e^{i(2\beta_a - G)z} - i\kappa_{ab} B_1 e^{i(\beta_a - \beta_b)z} \\ \frac{\partial A_2}{\partial z} &= i\kappa_{aa} A_1 e^{-i(2\beta_a - G)z} + i\kappa_{ab} B_2 e^{i(\beta_b - \beta_a)z} \\ \frac{\partial B_1}{\partial z} &= -i\kappa_{ba} e^{iGz} A_1 e^{-i(\beta_a - \beta_b)z} - i\kappa_{bb} B_1 - i\kappa_{bb} B_2 e^{2i\beta_b z} \\ \frac{\partial B_2}{\partial z} &= i\kappa_{ba} e^{-iGz} A_2 e^{i(\beta_a - \beta_b)z} + i\kappa_{bb} B_1 e^{-i2\beta_b z} + i\kappa_{bb} B_2 \end{aligned} \quad (5)$$

where

$$\begin{aligned} \kappa_{ja} &= n_s (\omega/4) \int e_{yj}^* n_{1b} e_{ya} dx \\ \kappa_{jb} &= n_s (\omega/2) \int e_{yj}^* n_{1a} e_{yb} dx; \quad j = a, b \end{aligned} \quad (6)$$

are the coupling constants between the corresponding modes. For simplicity, we assume the following relationships:

$$\begin{aligned} a_1(z) &= A_1(z) e^{-i\Delta\beta_a z} \\ a_2(z) &= A_2(z) e^{i\Delta\beta_a z} \\ a_3(z) &= B_1 e^{-i(\Delta\beta_b - \kappa_{bb})z} \\ a_4(z) &= B_2 e^{i(\Delta\beta_b - \kappa_{bb})z} \end{aligned} \quad (7)$$

where

$$\begin{aligned}
2\Delta\beta_a &= 2\beta_a - G \\
2\Delta\beta_b &= 2\beta_b + 2\kappa_{bb} - G
\end{aligned} \tag{8}$$

The coupled-mode equations (5) can be presented in a matrix form:

$$\frac{\partial}{\partial z} \begin{pmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{pmatrix} = \begin{pmatrix} -i\Delta\beta_a & -i\kappa_{aa} & -i\kappa_{ab} & 0 \\ i\kappa_{aa} & i\Delta\beta_a & 0 & i\kappa_{ab} \\ -i\kappa_{ba}e^{iGz} & 0 & -i\Delta\beta_b & -i\kappa_{bb}e^{iGz} \\ 0 & i\kappa_{ba}e^{-iGz} & i\kappa_{bb}e^{-iGz} & i\Delta\beta_b \end{pmatrix} \begin{pmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{pmatrix} \tag{9}$$

In order to solve Equation (9), one can assume solutions (Zill, 1982): $\tilde{a} = \tilde{\alpha}e^{sz}$, where s is the propagation factor or eigenvalue of the operator $\partial/\partial z$ with eigenmode \tilde{a} , and $\tilde{\alpha}$ is the spatial structure. The substitution of this form to Equation (9) will result in the following secular equation

$$\begin{pmatrix} -i\Delta\beta_a - s & -i\kappa_{aa} & -i\kappa_{ab} & 0 \\ i\kappa_{aa} & i\Delta\beta_a - s & 0 & i\kappa_{ab} \\ -i\kappa_{ba}e^{iGz} & 0 & -i\Delta\beta_b - s & -i\kappa_{bb}e^{iGz} \\ 0 & i\kappa_{ba}e^{-iGz} & i\kappa_{bb}e^{-iGz} & i\Delta\beta_b - s \end{pmatrix} \begin{pmatrix} \alpha_1 \\ \alpha_2 \\ \alpha_3 \\ \alpha_4 \end{pmatrix} = 0 \tag{10}$$

The eigenvalue of $\{s_i\}$ can be found as the roots of the matrix determinant of Equation (10), i.e., for the special case, $\kappa_{aa}=\kappa_{bb}=\kappa$, and $\kappa_{ba}=\kappa_{ab}$ and $\Delta\beta_a=\Delta\beta_b=\Delta\beta$ ($\beta \gg \kappa_{bb}$),

$$s_{1,\dots,4} = \pm \left\{ \kappa^2 - \Delta\beta^2 \pm \kappa_{ab} \sqrt{2\kappa^2 - \kappa_{ab}^2} \right\}^{1/2} \tag{11}$$

The spatial structures $\{\alpha_{ni}\}$ can be found after the substitution of eigenvalues $\{s_i\}$ into Equation (10), resulting:

$$\begin{aligned}
\alpha_{2i} &= \frac{-(i\Delta\beta_a + s_i)(\Delta\beta_b^2 + s_i^2 + \kappa^2)}{i\kappa(\Delta\beta_b^2 + s_i^2 + \kappa^2 + \kappa_{ab}^2)} \alpha_{1i} \\
\alpha_{3i} &= \frac{(-i\Delta\beta_a - s_i)}{i\kappa_{ab}} \alpha_{1i} - \frac{\kappa}{\kappa_{ab}} \alpha_{2i} \\
\alpha_{4i} &= -\frac{\kappa}{\kappa_{ab}} \alpha_{1i} - \frac{(i\Delta\beta_a - s_i)}{i\kappa_{ab}} \alpha_{2i}
\end{aligned} \tag{12}$$

where $\{\alpha_{1i}\}$ may be chosen as constants. Using the eigenvalues and spatial structures, now we have eigenmodes $\{a_n\}$ as:

$$a_n = \sum_i C_i \alpha_{ni} e^{s_i z} \tag{13}$$

where $\{C_i\}$ are constants that have to be determined by using boundary conditions.

Now, from Equations (13) and (12), we obtain the modes as follows:

$$\begin{aligned}
A_1(z) &= [C_1 \alpha_{11} e^{s_1 z} + C_2 \alpha_{12} e^{s_2 z} + C_3 \alpha_{13} e^{s_3 z} + C_4 \alpha_{14} e^{s_4 z}] e^{-i\Delta\beta z} \\
A_2(z) &= [C_1 \alpha_{21} e^{s_1(z-L)} + C_2 \alpha_{22} e^{s_2(z-L)} + C_3 \alpha_{23} e^{s_3(z-L)} + C_4 \alpha_{24} e^{s_4(z-L)}] e^{i\Delta\beta(z-L)} \\
B_1(z) &= [C_1 \alpha_{31} e^{s_1 z} + C_2 \alpha_{32} e^{s_2 z} + C_3 \alpha_{33} e^{s_3 z} + C_4 \alpha_{34} e^{s_4 z}] e^{-i\Delta\beta z} \\
B_2(z) &= [C_1 \alpha_{41} e^{s_1(z-L)} + C_2 \alpha_{42} e^{s_2(z-L)} + C_3 \alpha_{43} e^{s_3(z-L)} + C_4 \alpha_{44} e^{s_4(z-L)}] e^{i\Delta\beta(z-L)}
\end{aligned} \tag{14}$$

In order to determine $\{C_i\}$, we apply boundary conditions; at $z = 0$, $A_1(0) = A_o$, $B_1(0) = 0$, and at $z = L$, $A_2(L) = B_2(L) = 0$. Therefore, from Equation (14) we obtain :

$$\sum_i C_i \alpha_{1i} = A_o; \quad \sum_i C_i \alpha_{2i} = 0; \quad \sum_i C_i \alpha_{3i} = 0; \quad \sum_i C_i \alpha_{4i} = 0 \tag{15}$$

By substituting Equations (11) and (12) into Equation (14), and employing the Cramer rule (Boas, 19843), we may find the values of $\{C_i\}$ that are needed in Equation (14).

Finally, the device parameters, drop output, return loss and total transmission can be calculated numerically by applying boundary condition at $z = 0$ as follows:

$$\text{Drop output} = |B_2(0)|^2 / |A_o|^2 \tag{16}$$

$$\text{Return loss} = |A_2(0)|^2 / |A_o|^2 \quad (17)$$

$$\text{Total transmission} = 1 - \left[|B_2(0)|^2 + |A_2(0)|^2 \right] / |A_o|^2 \quad (18)$$

where $A_o = A(0)$.

RESULTS AND ANALYSIS

First, we check the reflectance of a single Bragg grating that we will use later in the add drop filter. By taking the parameters $L = 1$ cm, $n_s = 1.5$, and $\kappa = 5/\text{cm}$, the reflectance as a function of $\Delta\lambda$ ($= \lambda - \lambda_B$) is shown in Figure 2. It shows that the central reflectance is 100% with a bandwidth of 0.3 nm around the Bragg wavelength $\lambda_B = 1550$ nm (Yeh, 1988).

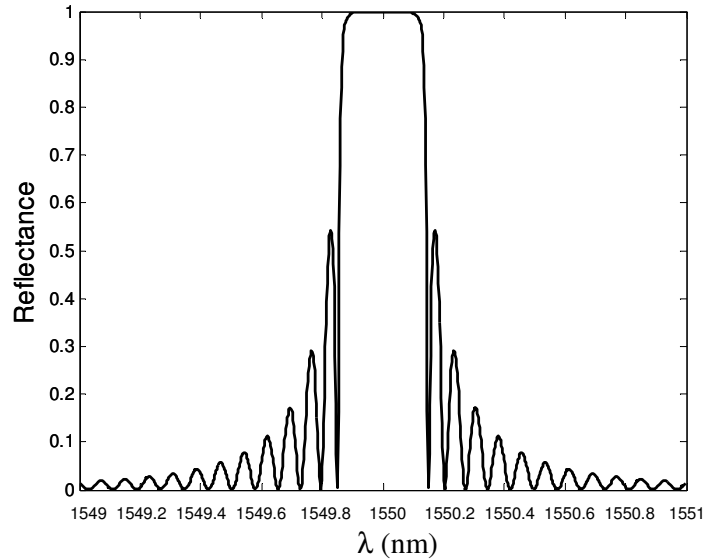


Figure 2. Reflectance of a Bragg grating with $L = 1$ cm, $n_s = 1.5$, and $\kappa = 5 \text{ cm}^{-1}$.

Second, the device parameters shown by Equations (16-18) are calculated using the following values: $L = 1$ cm, $n_s = 1.5$, $\lambda_B = 1.550$ μm , $\kappa = 5$ cm^{-1} , and $\kappa_{ab} = 0.5/\text{cm}$. The drop output as a function of $\Delta\lambda$ is shown by Figure 3. The drop output of 80% at the Bragg wavelength with a bandwidth of 0.1 nm or 12.5 GHz is observed. It shows also that there is no drop output found outside the bandwidth. Although the width at the base of the drop output (= 3 nm) is equal with that of reflectance in Figure 2, it becomes much smaller at the half peak (= 1 nm) and no side lobe is observed. This result indicates that these parameters are good parameters for an effective add-drop filter.

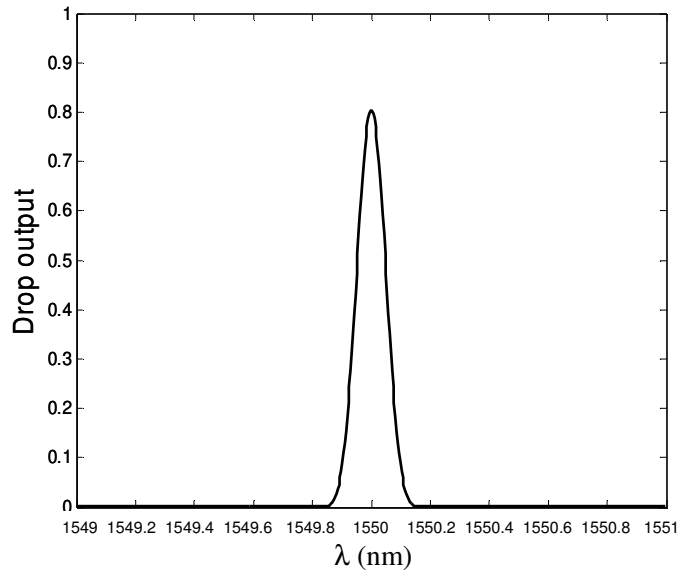


Figure 3. Drop output of the device as a function of $\Delta\lambda$. The input parameters are $L =$

1 cm, $n_s = 1.5$, $\lambda_B = 1.550$ μm , $\kappa = 5$ cm^{-1} , and $\kappa_{ab} = 0.5$ cm^{-1} .

The return loss is shown in Figure 4; with the same bandwidth as shown in Figure 3. The return loss is very small, i.e. 0.02% at the Bragg wavelength which indicates that only a small fraction of light is reflected by the device. Figure 5 shows the total transmission of the device. It is clear that the total transmission of the Bragg wavelength is large. From the combination of the high drop output, very small of return loss and narrow return loss, this type of filter satisfies the requirements for wavelength-division multiplexer applications. Beside that, the process of the drop and add is very fast because the coupling between modes occurs without the help from outside the filter.

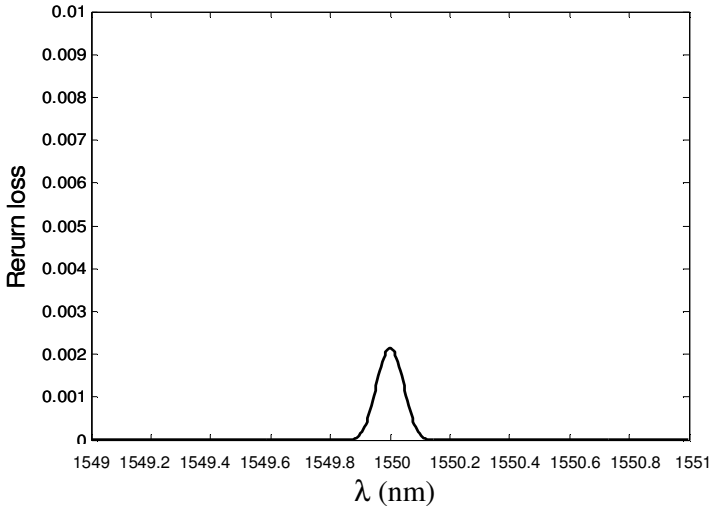


Figure 4. Return loss as a function of $\Delta\lambda$; $L=1\text{cm}$, $\kappa=5\text{ cm}^{-1}$, and $\kappa_{\text{ab}}=0.5\text{ cm}^{-1}$.

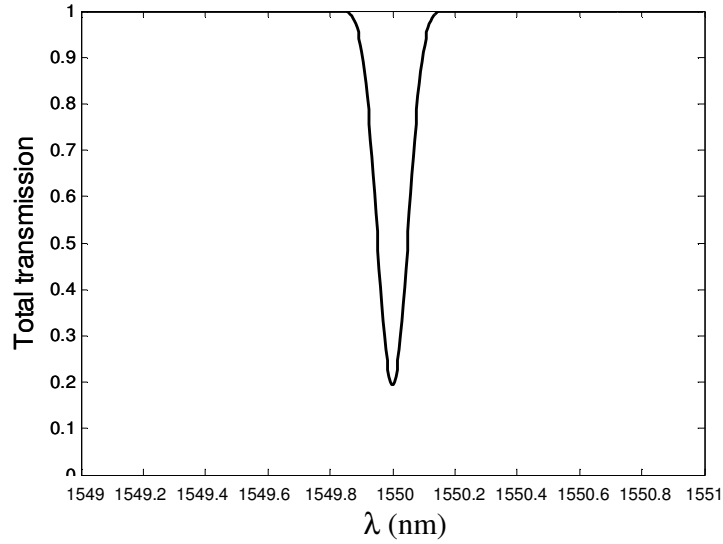


Figure 5. Total transmission as a function of $\Delta\lambda$; $L=1\text{cm}$, $\kappa=5\text{ cm}^{-1}$, and $\kappa_{ab}=0.5\text{ cm}^{-1}$.

Finally, we studied the influence of the coupling between two co-propagating modes, κ_{ab} , on the drop output by calculating the drop output as a function of κ_{ab} for $\kappa=5/\text{cm}$ with $L=0.9, 0.95,$ and 1 cm , respectively. The results are shown in Figure 6, where the drop output for each length increases with κ_{ab} . The value of κ_{ab} depends on the overlap integral between transverse fields, $e_{ya}(x)$ and $e_{yb}(x)$ in a waveguide, as indicated by Equation (6). If the separation is made smaller, the overlap and the coupling κ_{ab} will increase. We see also that the drop output for each value of κ_{ab} is increased for longer filter. It means, the power transfer is more effective for longer filter. Furthermore, for a value of drop output, shorter length needs bigger κ_{ab} or narrower separation between waveguides.

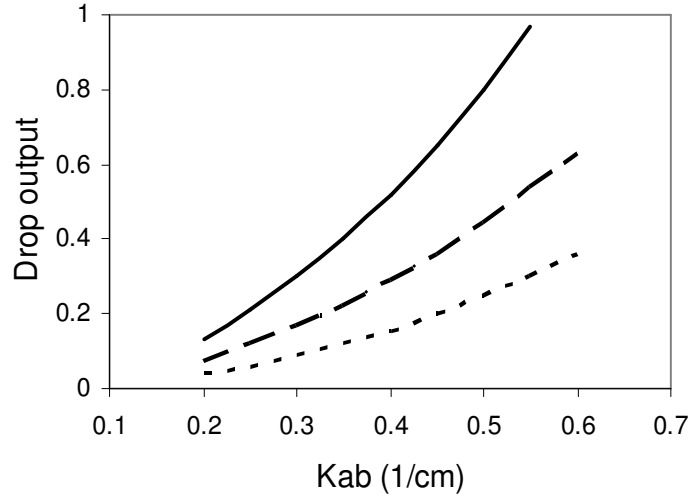


Figure 6. Drop output as a function of κ_{ab} for $\kappa = 5 \text{ cm}^{-1}$, $L = 0.9 \text{ cm}$ (---), $L = 0.95 \text{ cm}$ (---), and $L = 1 \text{ cm}$ (—).

CONCLUSION

We have studied an add-drop filter using a planar directional coupler with a single Bragg grating in its arms with a Bragg wavelength of $1.550 \mu\text{m}$. After we solved the coupled-mode equations, the filter parameters were calculated. The results are: for $\kappa = 5 \text{ cm}^{-1}$, $\kappa_{ab} = 0.5 \text{ cm}^{-1}$ and $L = 1 \text{ cm}$, (i) the drop output is 80% while the return loss $\approx 2.5 \times 10^{-3}$ of drop output; the bandwidth is very narrow, i.e., 0.1 nm , (ii) the drop output increase with the co-propagating coupling κ_{ab} and, (iii) the drop output increase with the filter length. These results are the characteristic of an add-drop filter that satisfy the requirements for wavelength-division multiplexing (WDM) applications.

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